

Low-noise Receiver / Scope Combinations For PDV

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This work was done by National Security Technologies, LLC, under Contract No. DE-AC52-06NA25946 with the U.S. Department of Energy and supported by the Site-Directed Research and Development Program.



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Background

- Currently using high-gain receivers with fast digitizers
- With new generation of digitizers with useful sensitivity below 20 mV/div, there is an opportunity to use lower-gain receivers
- What PDV requires is measurement of multiple frequencies across a very wide variety of signal levels: High dynamic range
- What PDV does not require:
 - accurate phase across long records
 - exquisitely low harmonic distortion
- We will use SNR as a means to estimate the frequency-domain effective number of bits (“FNOB”) across a wide array of input signal levels



SNR relationship to ENOB

$$\text{SNR}_{\text{TIME-DOMAIN,dB}} = 6.02 \cdot \text{ENOB} + 1.76 + 20 \log (2 \cdot \text{RMS} / V_{\text{full-scale}})$$

ENOB = effective bits for digitizer, V = full scale range, A = RMS amplitude of applied signal
See Wiley Encyclopedia of Electrical and Electronics Engineering, Vol. 18, J. Blair

$$\text{SNR}_{f,\text{dB}} = 6.02 \cdot \text{ENOB} + 1.76 + 20 \cdot \log(2 \cdot \text{RMS} / V_{\text{FS}}) + 10 \cdot \log(N_{\text{FFT}}/2)$$

Normalize for fraction of full-scale used

SNR increase by using frequency-domain analysis

Number of bits as function of SNR, fraction of full-scale, FFT RBW

"Frequency-domain Number of Bits"

$$\text{"FNOB"} = (1/6.02) \cdot (\text{SNR}_{f,\text{dB}} - 1.76 - 20 \cdot \log(2 \cdot \text{RMS} / V_{\text{FS}}) - 10 \cdot \log(N_{\text{FFT}}/2))$$

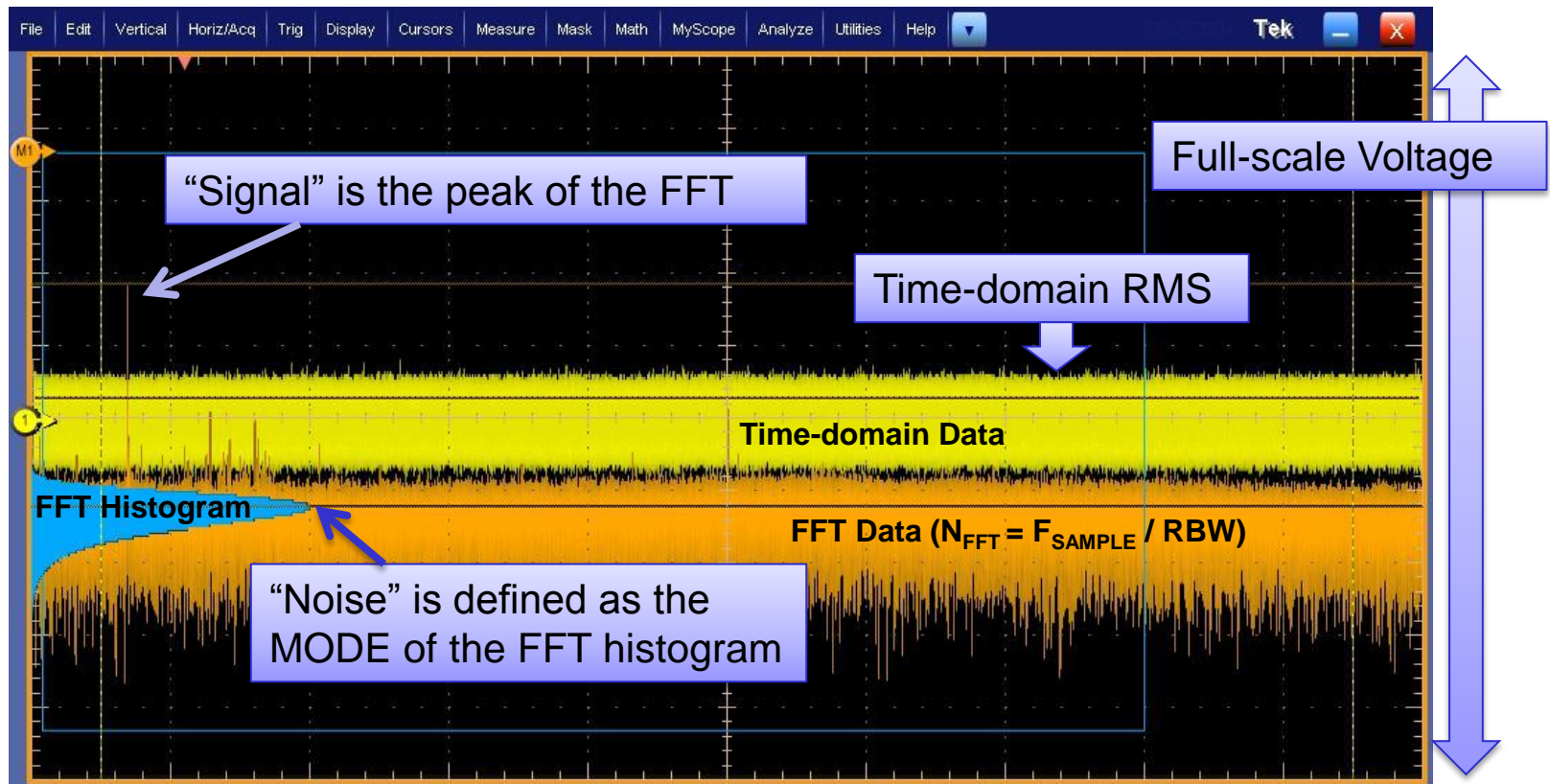
Measure this

Measure this

Fix this using Resolution Bandwidth (RBW) for FFT



On-board analysis for Lab data



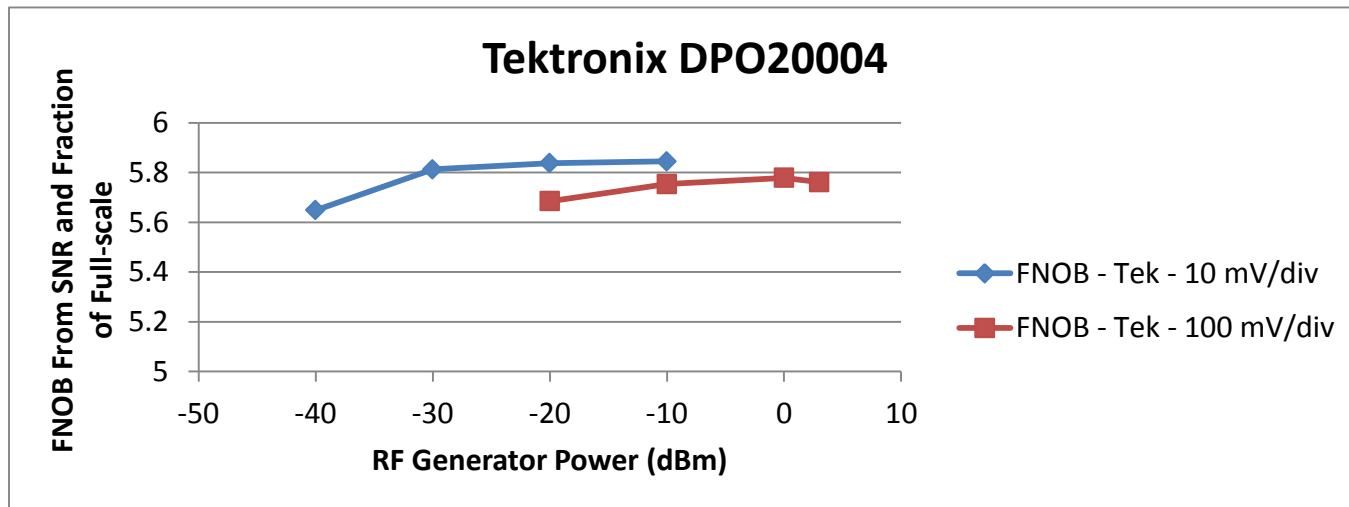
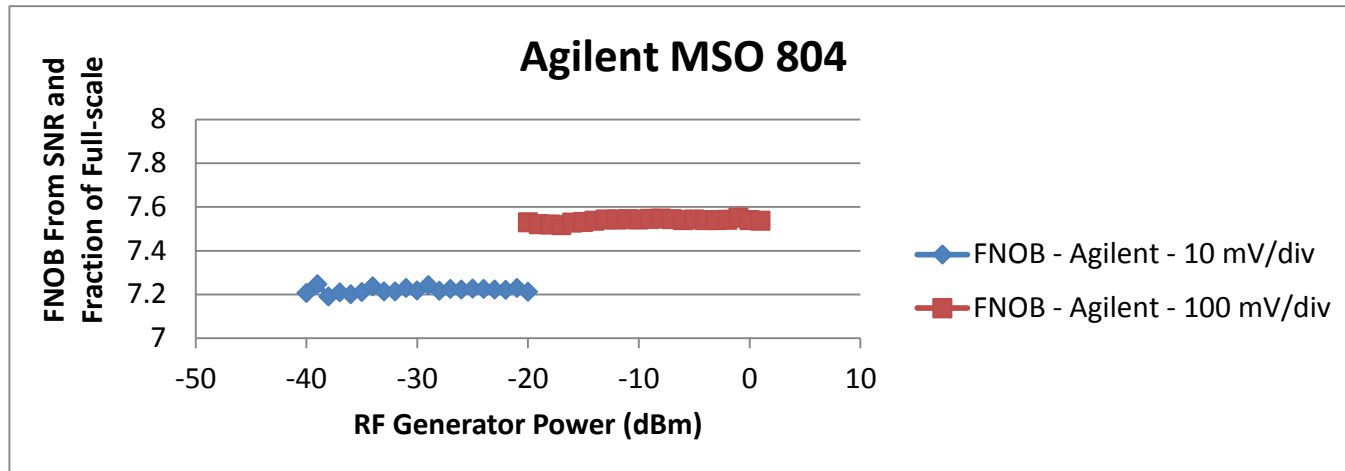
$$\text{"FNOB"} = (1/6.02) * (\text{SNR}_{f, \text{dB}} - 1.76 - 20 * \log(2 * \text{RMS} / V_{FS}) - 10 * \log(N_{FFT} / 2))$$

Lab Measurements: Scopes and receivers

- Purely electrical:
 - Scope + RF Generator: SNR at select sensitivities → FNOB
- Optical and electrical
 - Scope + Receiver dark-noise vs. Sensitivity
 - This returns the setting below which there will be no improvement in SNR
 - Comparison with 50 Ω Termination
 - Scope + Receiver + 1 GHz interference from two RIO ORION lasers
 - SNR: Adjust LO level at fixed “signal” power
 - SNR: Adjust “signal” power at fixed LO power
- By measuring SNR and fraction of scope full-scale, we can calculate the frequency-domain number of bits, “FNOB”



Scope comparison: Electrical (1.05 GHz)

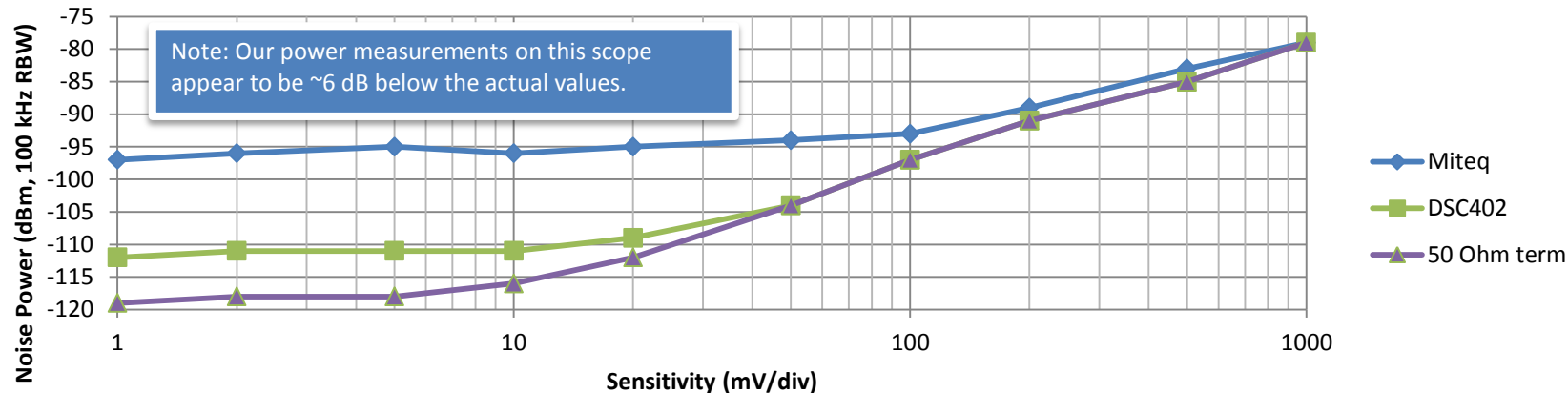


$$\text{"FNOB"} = (1/6.02) * (\text{SNR}_{f,\text{dB}} - 1.76 - 20 * \log(2 * \text{RMS}/V_{\text{FS}}) - 10 * \log(N_{\text{FFT}}/2))$$

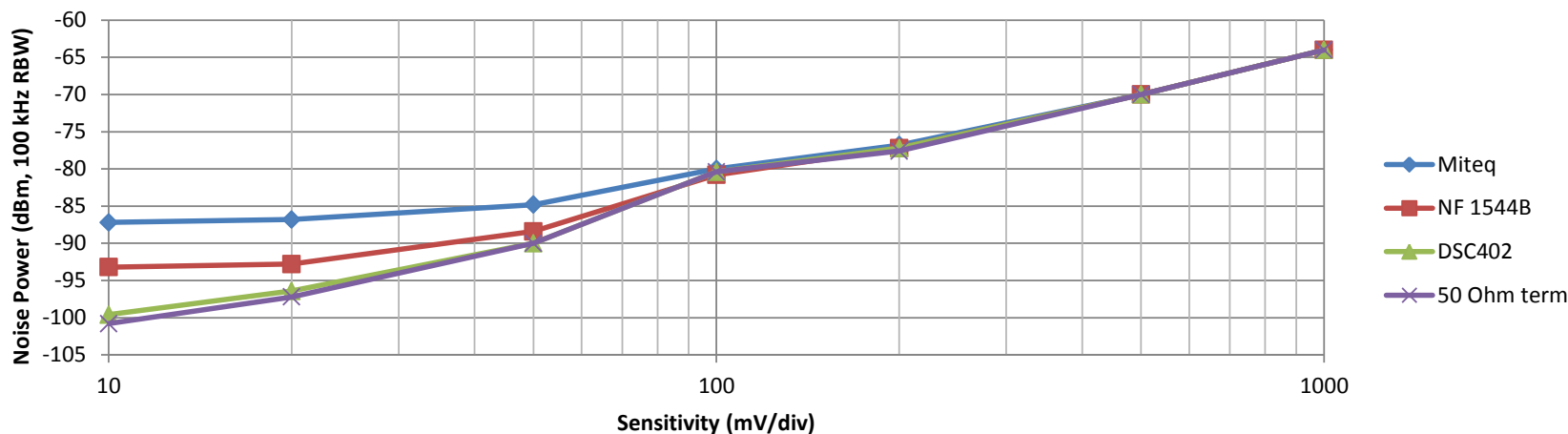


Scope Comparison: Optical Receiver Dark Noise

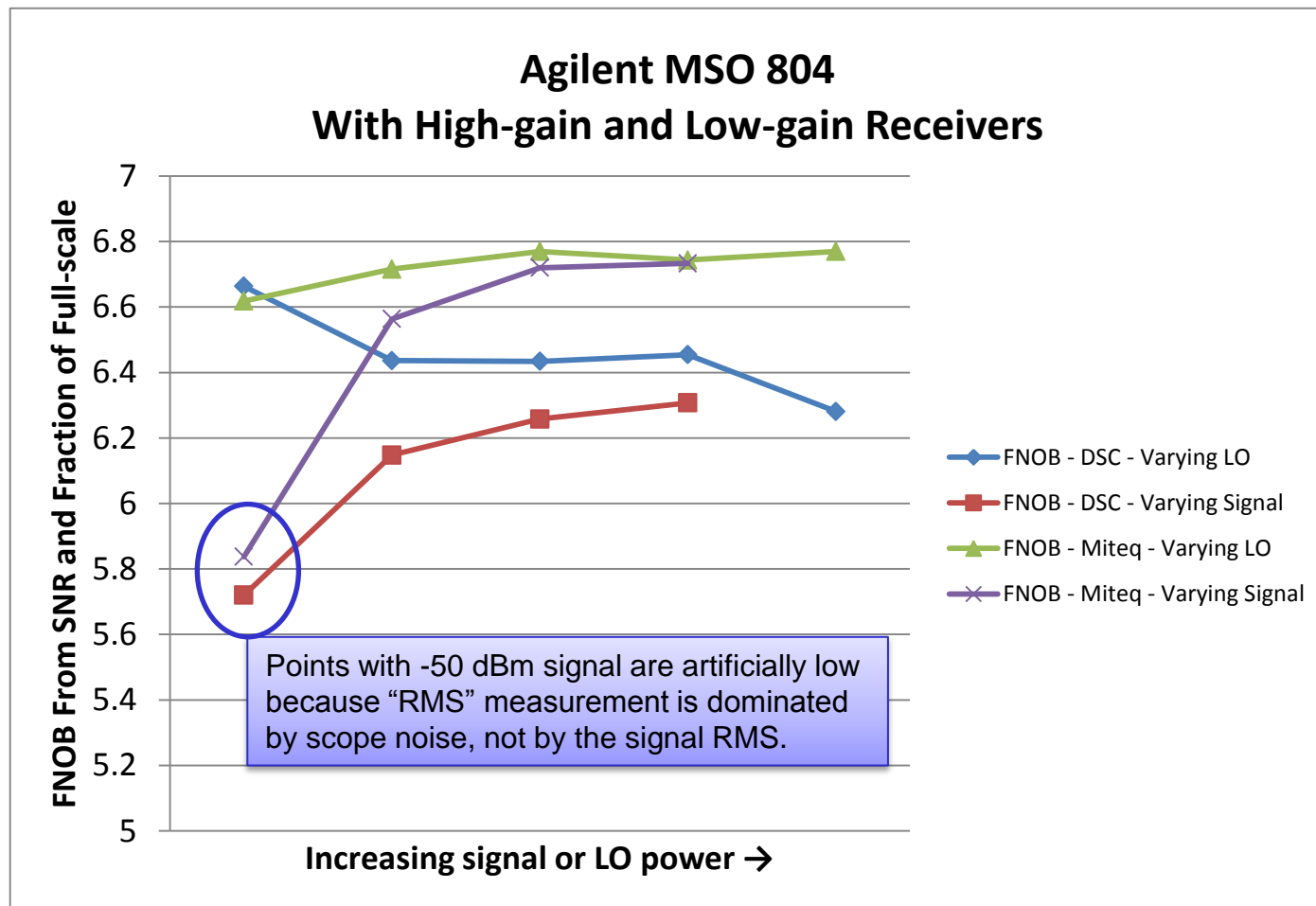
Agilent MSO 804 / Receiver Dark Noise



Tektronix DPO20004 / Receiver Dark Noise



Lab Data: Optical signals on the Agilent MSO 804



Varying LO: -20 dBm to 0 dBm in steps of 5 dBm; Signal = -20 dBm
 Varying Signal: -50 dBm to -20 dBm in steps of 10 dBm; LO = 0 dBm

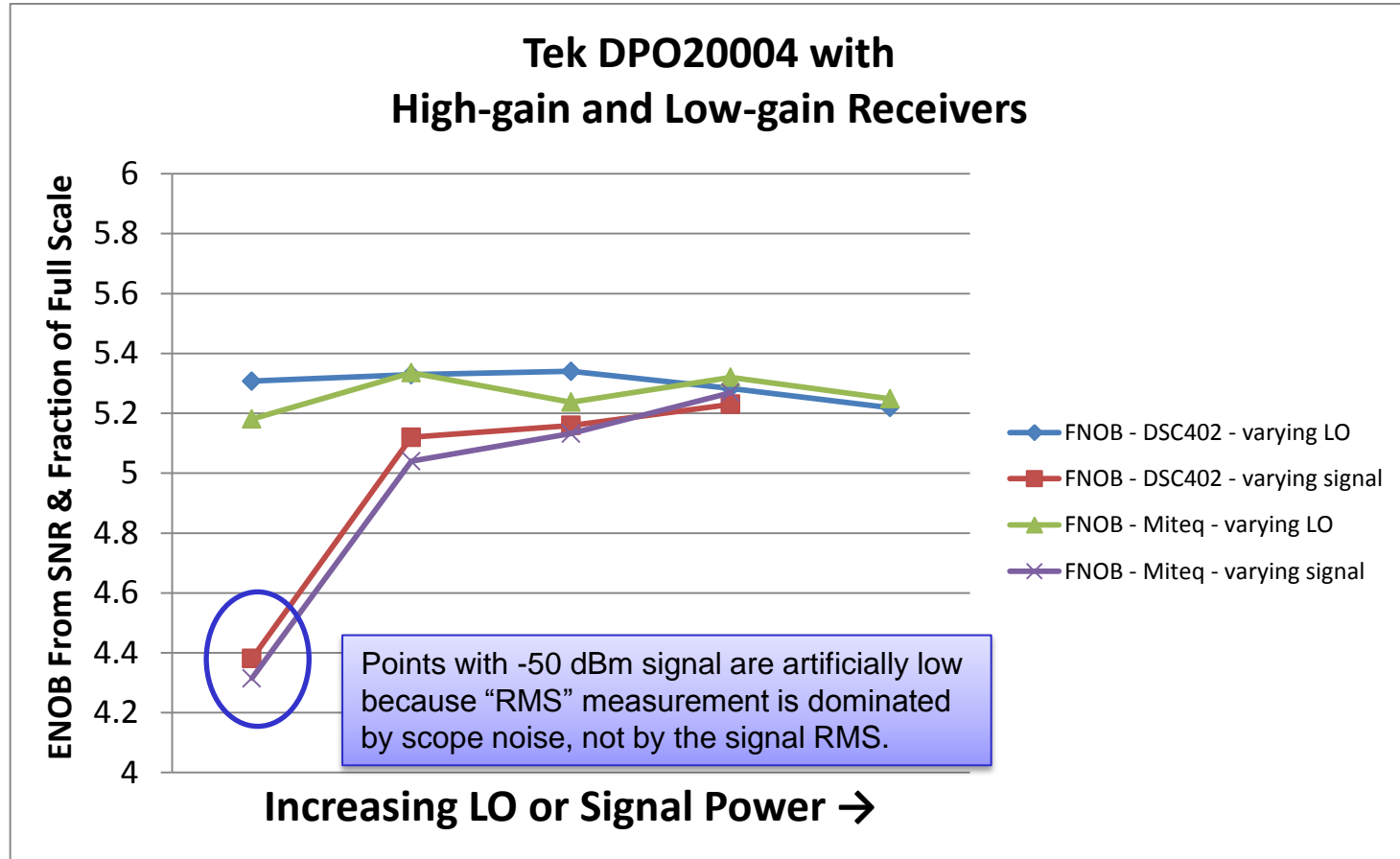


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Lab Data: Optical signals on the Tektronix DPO20004



Varying LO: -20 dBm to 0 dBm in steps of 5 dBm; Signal = -20 dBm
 Varying Signal: -50 dBm to -20 dBm in steps of 10 dBm; LO = 0 dBm



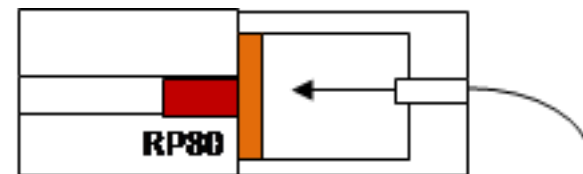
Lab data: Conclusions

LAB DATA	Electrical FNOB	Optical FNOB
Agilent MSO 804	7.4 ± 0.2	6.5 ± 0.3
Tek DPO20004	5.8 ± 0.1	5.25 ± 0.1

- Idealized conditions:
 - No circulators
 - No long transmission fibers
 - No back-reflections
- Agilent MSO 804 seems to give higher dynamic range
 - Can we exploit this on shot data?
- Comparable performance with both receiver configurations
 - Discovery 402 (0.35 V/W) at 10 mV/div
 - Miteq DR125G (2 V/W) at 100 mV/div



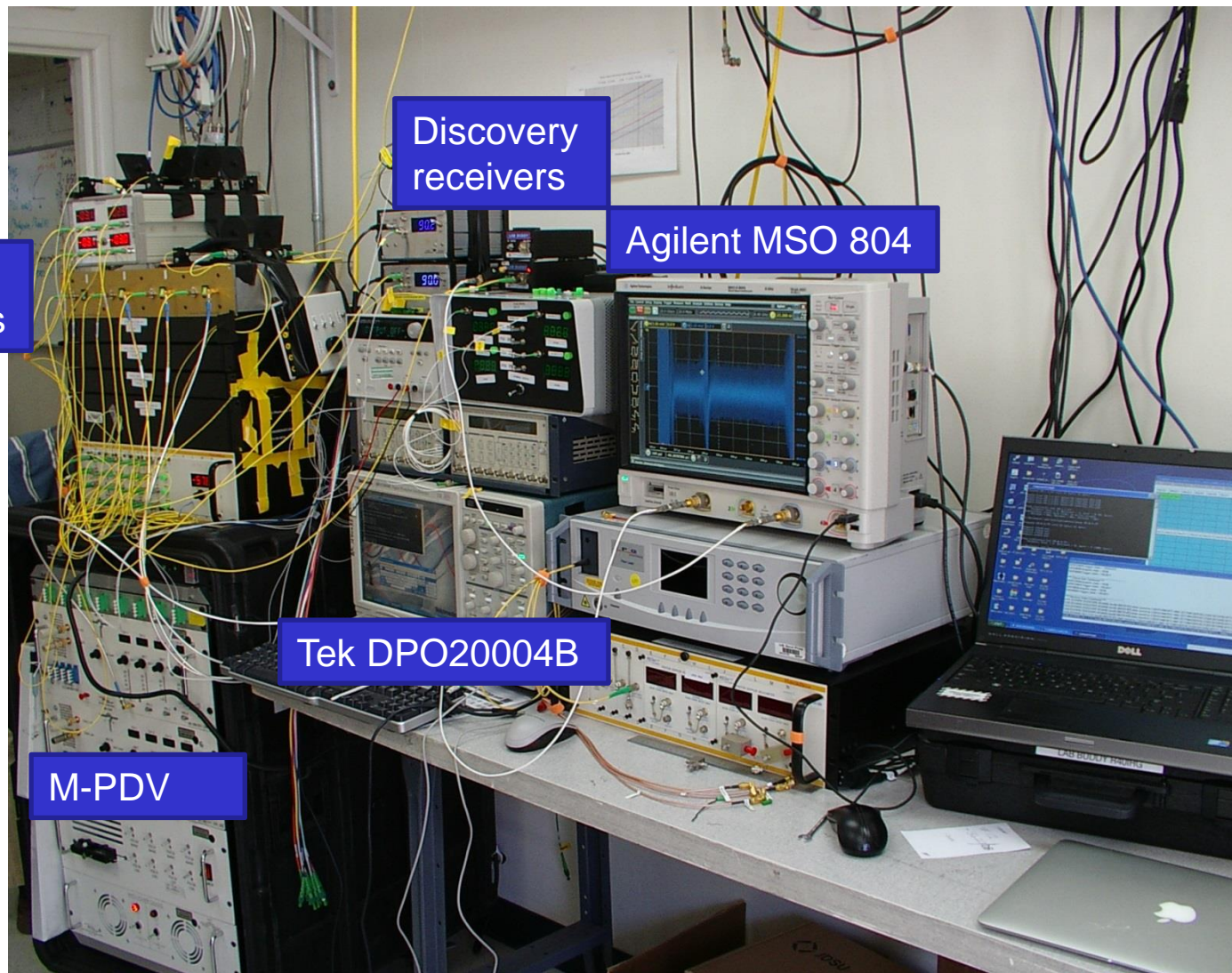
Shot data



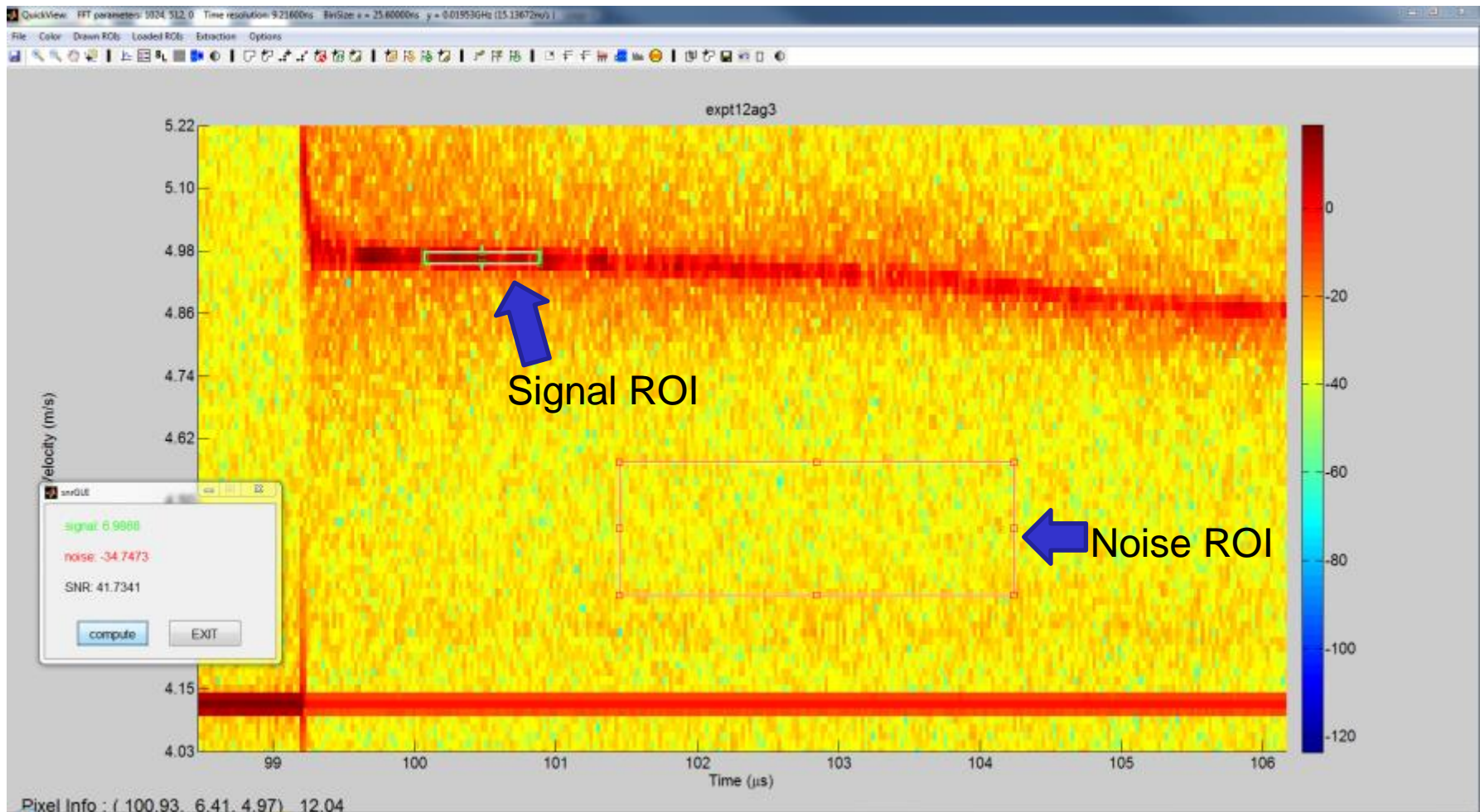
- STL BoomBox
- RP-80 detonators driving 1" diameter copper coupons
- Single PDV probe with multiple wavelengths, demux on back-end
- Velocities ~ 800 m/s
- All systems are heterodyne, zero-velocity frequencies from 1 – 6 GHz
- Variety of recording setups were tested
- Analysis using “Quick-view” with FFT window of 1024 points
- Two scopes used:
 - Tek DPO72004B (run at 25 GS/s, 12.5 GHz)
 - Agilent MSO 804 (run at 20 GS/s, 8.4 GHz)
- Two receivers used:
 - Miteq DR125G (12.5 GHz), ~2 V/mW
 - Discovery DSC402, ~0.35 V/mW



Recording setup

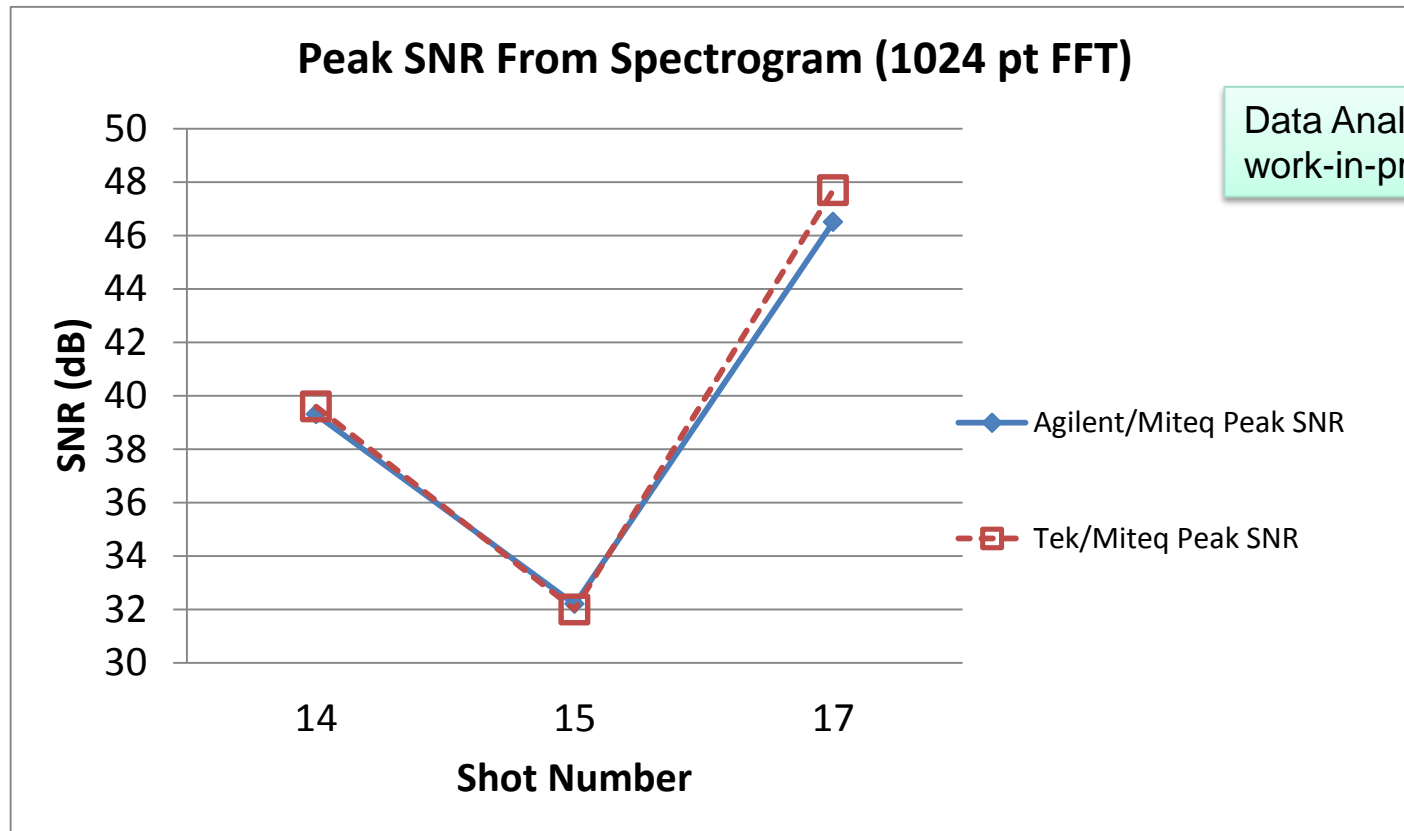
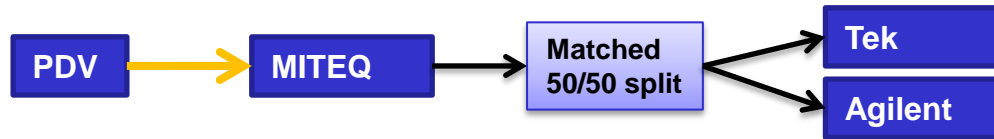


Typical Spectrogram and Analysis

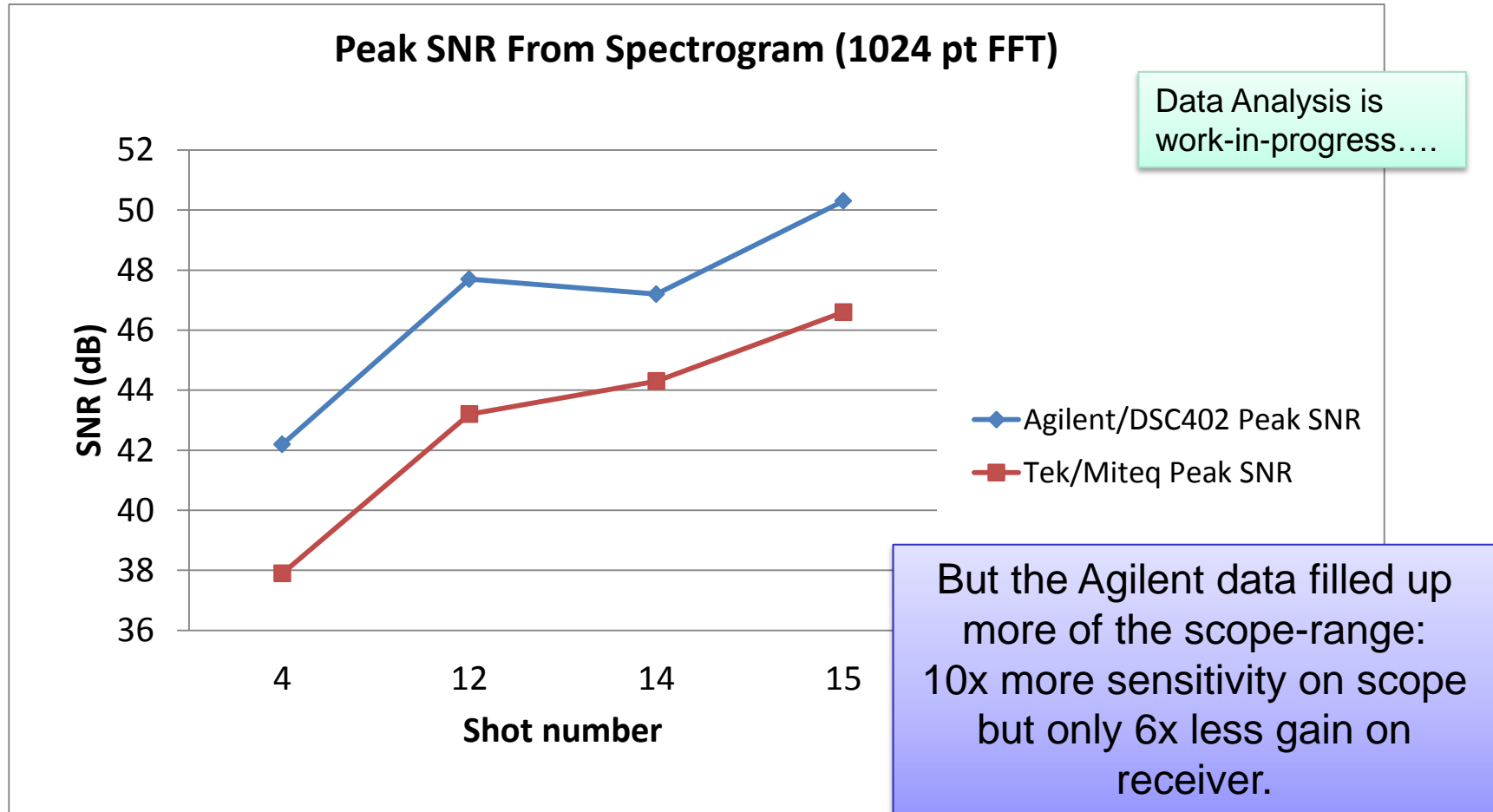
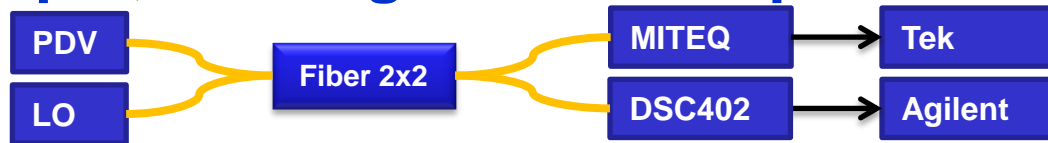


Shot data:

Electrical split, Miteq, Agilent and Tek, both on 100 mV/div



Shot data: Optical split, DSC/Agilent & Miteq/Tek



Normalize for scope range

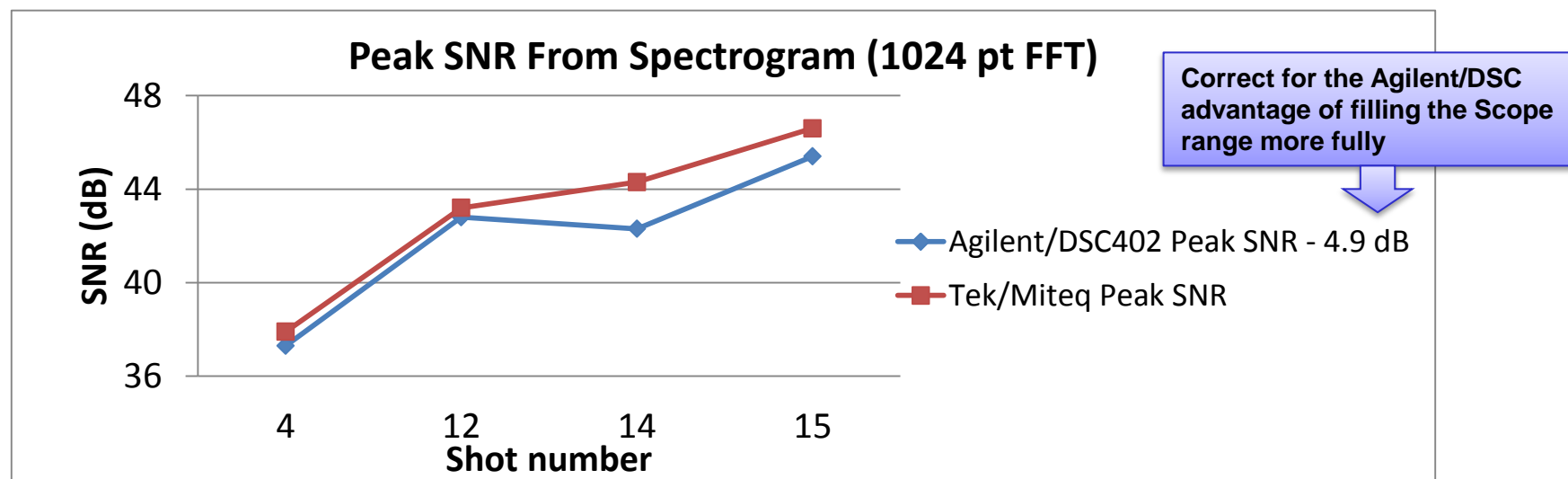
$$\text{SNR}_{f,\text{dB}} = 6.02 \cdot \text{FNOB} + 1.76 + 20 \cdot \log(2 \cdot \text{RMS} / V_{\text{FS}}) + 10 \cdot \log(N_{\text{FFT}} / 2)$$

This will be proportional to
(Receiver Gain) / (Scope Sensitivity)

DSC / Agilent: $20 \log(0.35 \text{ V/mW} / 0.1 \text{ mV}) = 10.9$

Miteq / Tek: $20 \log(2.00 \text{ V/mW} / 1.0 \text{ V}) = 6.02$

So, just based on fraction of full-scale, the DSC/Agilent setup is getting a bonus 4.9 dB of SNR
Correct for coverage difference, and the result will be purely the “FNOB” difference



Conclusions

- Shot data:
 - No clear difference on coverage-normalized shot data
 - With small signals, the DSC receiver will fill up more of the full-scale range at 10 mV/div... potential for better SNR
- Outstanding questions
 - Why does the dynamic range on actual shot data seem to be limited by the PDV hardware?
 - Circulator leakage?
 - Brillouin backscatter?
 - Something else?
 - What about the tin shots, where there was no distinct surface?
 - What conditions (system design, experiment...) would allow us to exploit the extra dynamic range we see in the lab?
- Shot-data experiments I'd like to do in the future:
 - DSC receiver vs Miteq receiver on the same scope
 - Homodyne PDV – DC-block the zero-velocity signal
 - Put circulator near probe to reduce Rayleigh backscatter



Thank you

- Agilent for loaning the MSO 804 for this work
- NSTec Program Support
 - Detectors & Instrumentation Project
 - Shock-wave R&D Project
- LANL team
- NSTec STL team
- NSTec NLV team
- All the vendors who are making enabling components for PDV systems!
- *And of course, to Ed and Larry for making this workshop possible!*

